

### EMF and Termínal Voltage

**Example 1: Consider** the circuit to the right. If the resistors represent light bulbs:

a.) What does the ammeter read when the switch is open? (step 1—redraw without the meters) All the battery's voltage drop happens across the resistor  $R_1$ , and the current through the ammeter is just  $i_1$ , so:

$$V_{bat} = i_1 R_1 = i_0 R_1 \implies i_0 = \frac{V_{bat}}{R_1}$$

b.) In an ideal world, what should happen to the current through  $\mathbf{R}_1$  when the switch is thrown?



Nothing, and this is tricky. If you think of the current from the battery as fixed, you'll conclude  $R_1$ 's current will drop as some current will now be needed for  $R_2$ . But au, contraire. Current through  $R_1$  is governed by the voltage across it and its resistance, neither of which changes when you throw the switch. So what happens? The battery outputs MORE CURRENT so  $i_0 = i_1 + i_2$  and  $R_1$ , in theory, should remain unaffected.

c.) Here's the rub: if you carry this experiment out in the real world, the light bulb associated with  $R_1$ will *actually dim* suggesting that the current through  $R_1$  has diminished. So what's going on?

*The problem lies* in the internal workings of power supplies. There are actually two parts to a p.s.:

1.) *There is* the part of the battery that creates the electric field that motivates charge to move through the wires. This part has the units of *volts* and is called *the electromotive force,* or EMF (symbol  $\varepsilon$ );



Including the voltage drop due to  $r_i$ , this means a VOLTMETER will read what is called *the terminal voltage* (i.e., the voltage as measured at the terminals of the p.s.) equal to:  $V_{terminal} = \varepsilon - i_o r_i$ 

Soooo, if you increase  $i_0$  by throwing the switch,  $V_{terminal}$  does DOWN across the resistors and the bulbs will dim!

 $\mathbf{R}_{2}$ 

power supply

switch

### Series Resistor Combinations

**Example 2: Derive** an expression for the *equivalent resistance* (the single resistor that can take the place of the resistor combination) for the series combination shown to the right. Assume an ideal power supply with no internal resistance.

*The idea* behind  $R_{equ}$  is to find the single resistor that can take the place of all the resistors in the system. In other words, the single resistor that, when put across  $V_0$  will draw  $i_0$ 

*<sup>i</sup>Using the idea* that the sum of the voltage drops across all the resistors will equal the voltage drop across the power supply, and including Ohm's Law in the mix, we can write:

$$V_{o} = \Delta V_{i} + \Delta V_{2} + \Delta V_{3}$$

$$\dot{j}_{o}R_{eq} = \dot{j}_{o}R_{1} + \dot{j}_{o}R_{2} + \dot{j}_{o}R_{3}$$

$$\Rightarrow R_{eq} = R_{1} + R_{2} + R_{3}$$



## Characterístics of a Series Combinations

*--Each element* in a series combination is attached to its neighbor in *one place only*.

 $=18 \Omega$ 

*--Current* is common to each element in a series combination.

*--There are* no nodes (junctions—places where current can slit up) internal to series combinations.

--*The equivalent resistance* for a series combination is:  $R_{eq} = R_1 + R_2 + R_3 + ...$ 

*--This means* the equivalent resistance is larger than the largest resistor in the combination;

*--This means* that if you add a resistor to the combination,  $R_{eq}$  will *increase* and the current through the combination (for a given voltage) will *decrease*.

**Example 3:** What's the equivalent resistance of a 5  $\Omega$ , 6  $\Omega$  and 7  $\Omega$  resistor in series?  $R_{eq} = (5\Omega) + (6\Omega) + (7\Omega)$ 

### Parallel Resistor Combinations

**Example 3: Deríve** an expression for the *equivalent resistance* (the single resistor that can take the place of the resistor combination) for the parallel combination shown to the right. Assume an ideal power supply with no internal resistance.

*What's common* in a parallel combination is the voltage drop across each element.

*Also, in this case,* the sum of the currents through the parallel combination must equal the current drawn from the power supply. Using that and Ohm's Law, we can write:

$$i_{o} = i_{i} + i_{2} + i_{3}$$

$$\frac{\cancel{N}_{o}}{R_{eq}} = \frac{\cancel{N}_{o}}{R_{1}} + \frac{\cancel{N}_{o}}{R_{2}} + \frac{\cancel{N}_{o}}{R_{3}}$$

$$\implies \frac{1}{R_{eq}} = \frac{1}{R_{1}} + \frac{1}{R_{2}} + \frac{1}{R_{3}}$$

 $\begin{bmatrix} \mathbf{V}_{o} \\ \mathbf{R}_{equ} \\ \mathbf{V}_{o} \end{bmatrix}$ 

 $\mathbf{K}_{3}$ 

## Characterístics of a Parallel Combinations

--*Each element* in a series combination is attached to its neighbor in *two place*.

--Voltage is common to each element in a parallel combination.

*--There are* nodes (junctions—places where current can slit up) internal to parallel combinations.



--*The equivalent resistance* for a parallel combination  $\frac{1}{R_{eq}} = \frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} + \dots$ is:

*--This means* the equivalent resistance is SMALLER than the smallest resistor in the combination;

*--And*, if you add a resistor to the combination,  $R_{eq}$  will *decrease* and the current through the combination (for a given voltage) will *increase*.

*Example 3: What's* the equivalent resistance of three one-ohm resistors in parallel?

$$\frac{1}{R_{eq}} = \frac{1}{(1\Omega)} + \frac{1}{(1\Omega)} + \frac{1}{(1\Omega)}$$
$$\Rightarrow \frac{1}{R_{eq}} = 3 \Rightarrow R_{eq} = .333 \Omega$$

### Run and Shoot Problems

**Example 4:** An ideal power supply has  $EMF \epsilon = 10V$  powering it. Any blue letters that show up designate points on the circuit. Assume the *low voltage terminal* of the p.s. is at zero volts. Assume the resistor values are the same as the resistor subscripts.



a.) What is the first thing you would do if asked to work with this circuit? *Redraw the circuit* without the meters. They aren't doing anything in the circuit except identifying a branch or resistor for which you want a current.

b.) What is the absolute electrical potential at Point a? *There is essentially* no resistance between Point a and the high voltage terminal of the p.s., so their voltages are the same point being V<sub>a</sub> = 10v.
c.) What will the ammeter read?

*Some amount of* current is being drawn from the power supply. Being in the same branch as *Points a*, *b* and *g*, it will be the same for all three points. It will also be the current through the ammeter. So how do we get that?

#### c.) ammeter?

*Here is the circuit* with the meters removed. The trick here is to find the equivalent resistance for the circuit, then use Ohm's Law.

*This circuit* is  $R_1$  in series with  $R_2$  in parallel with  $R_3$  and  $R_4$  in series. That is:

$$\mathbf{R}_{eq} = \mathbf{R}_{1} + \left(\frac{1}{\mathbf{R}_{2}} + \frac{1}{\mathbf{R}_{3} + \mathbf{R}_{4}}\right)^{-1}$$
$$= (1 \ \Omega) + \left(\frac{1}{(2 \ \Omega)} + \frac{1}{(3 \ \Omega) + (4 \ \Omega)}\right)^{-1}$$
$$= 2.56 \ \Omega$$



Using the  $R_{eq}$  circuit:  $V_o = i_o R_{eq} \implies i_o = \frac{V_o}{R_{eq}}$  $= \frac{(10 \text{ V})}{(2.56 \Omega)} = 3.9 \text{ A}$ 

9.)

d.) How much power does  $R_2$  dissipate?

*We know* the absolute electrical potential (the voltage) at *Point a* is 10 volts.

We know current goes from high voltage to *low voltage*, so the *voltage change* across  $\mathbf{R}_1$ must be a voltage DROP equal to:

$$\Delta V_1 = i_o R_1 = (3.9 \text{ A})(1 \Omega) = 3.9 \text{ V}$$



=(10 V)-(3.9 V)

 $\Rightarrow$  V<sub>b</sub> = 6.1 V (= V<sub>c</sub>)

*Logic dictates* that the absolute electrical potential  $V_{\rm b} = V_{\rm a} - \Delta V_{\rm 1}$ at *Point b is*:

Because the absolute electrical potential at Point d is zero, the voltage across  $R_2$  equals  $V_c = 6.1 V$ and:

$$V_2 = i_2 R_2$$
  

$$\Rightarrow (6.1 V) = i_2 (2 \Omega)$$
  

$$\Rightarrow i_2 = 3.05 A$$

The power dissipated by  $\mathbf{R}_2$  is, then:  $P_2 = (i_2)^2$  $=(3.05 \text{ A})^2(2 \Omega)$ = 18.6 W

*e.)* What does the voltmeter read?

You should begin to see a pattern here. Every question, whether it be asking for an ammeter reading or voltmeter reading or power calculation or current through an element or voltage cross an element, they all require you to determine the CURRENT



through the branch in which the element exists. That, in general, is what you will always be doing—trying to derive expressions for current values.

In this case, you could determine the current through the far-right branch (so you could use Ohm's Law on  $\mathbb{R}_3$  to get what the voltmeter would read) by using the same approach we used to get the current in the central branch in Part c (you'd just be using *Points e* and *f* instead of *Points c* and *d* in the process).

*Or* . . .

Look at node h

$$i_{o} = i_{2} + i_{3}$$
  

$$\Rightarrow i_{3} = i_{o} - i_{2}$$
  

$$= 3.9A - 3.05A$$
  

$$= .85A$$

So

$$i_{meter} = i_{3}R_{3}$$
  
= (.85A)(3  $\Omega$ )  
= 2.55 V

#### **Example 5:** A power supply with $20 \Omega$

of internal resistance is used to power a circuit. If the current through  $R_4$  is .23 amps, what is the current through  $R_1$ ?

Start with what is obvious.

*The current* through  $R_1$  in the bottom branch will equal all the currents in the parallel combination put together, or

 $i_1 = i_2 + i_3 + i_4$ 

*We know* the current through  $R_4$ (given) and  $R_2$  (same size resistance with same voltage across it), so all we need is the current through  $R_3$ .

*The voltage* across each of the parallel resistors is the same, and equal to:

The the current through  $\mathbf{R}_3$  is:  $\mathbf{V}_3 = \mathbf{i}_3 \ \mathbf{R}_3$  $(1.61 \text{ V}) = \mathbf{i}_3(5 \Omega)$ 

So: 
$$i_1 = .23A + .32A + .23A$$
  
= .78A  
(1.61 V) =  $i_3$   
 $\Rightarrow i_3 = .32$ 



 $V_4 = i_4 R_4$ = (.23A)(7  $\Omega$ ) = 1.61 V

A



If the upper branch has half the resistance of the lower branch, it should draw twice the current.

*With 3 amps* coming in, that means 2 amps should pass through the upper branch.



**Note:** AP questions often have easy, non-mathematical, use-your-head solutions like this. That is why I'm showing you screwball problems like this. We will get into a more formal approach for analyzing circuit problems shortly.



between *Points a* and *b*?

Assuming the voltage at Points *a* is zero (this is tricky as you don't know what is—you can define it anyway you want, though), the voltage changes will be due to the increase due to the battery in the right branch and the drop due to the 6 ohm resistor. That is:  $V_{ab} = 24 - (2.75 \text{ A})(6 \Omega)$ 

*This is just* the *current* through the 3 ohm resistor, or:  $V_{ab} = i_2 R_3$  *This is just* the *current* through the 3 ohm resistor, or:  $7.5V = i_2(6 \Omega)$  *d.) How is*  $i_1$  generated? *Each battery produces* an E-fld,  $\Rightarrow i_2 = 1.25 A$ which permeates the entire circuit. The fields superimpose on one another, creating a net field. That net field is what motivates charge to move in each branch.

= 75 V

Some Definitions

A branch: A section of a circuit in which the current is the same everywhere.

*--elements in series* are a part of a single branch (look at sketch).

*--in the circuit* to the right, there are three branches.



A node: A junction where current can split up or be added to.

--elements in parallel have nodes internal to the combination.

--in the circuit above, there are two nodes.

A loop: Any closed path inside a circuit.

*--in a circuit*, loops can be traverse in a clockwise or counterclockwise direction *circuit* above, there are three loops.

#### For Your Amusement

For the circuit to the
right:
 a.) How many branches are there?
 six
 b.) How many nodes are there?
 four
 c.) How many loops are there?
 seven



And that last little nubbin is supposed to be a tooth, cause this looks like a face to me!

## Kírchoff's Laws—the Formal Approach

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With the definitions under your belt, Kirchoff's Laws are simple (and you've been inadvertently using them in the seat-ofthe-pants evaluations). They are:

Kirchoff's First Law: The sum of the currents into a node equals the sum of the currents out of a node. Mathematically, this is written as:  $\sum_{i_{\text{into node}}} i_{i_{\text{out of node}}}$ *Example* from the circuit's Node  $i_0 = i_2 + i_3$ 

Kirchoff's Second Law: The sum of the voltage changes around a closed path (a loop) equals ZERO. Mathematically, this is written as:  $\sum \Delta V = 0$ 

*Examples:* starting at Node A:

Loop 1 traversing counterclockwise:

 $R_1i_0 - \varepsilon + R_2i_2 = 0$ 

Loop 2 traversing clockwise:

Loop 1

 $-R_{3}i_{3} - R_{4}i_{3} + R_{2}i_{2} = 0$ 

Node A

Loop 2

*Note:* Current moves from hi to lo voltage, so traversing against the current through a resistor produces a  $\Delta V$  that is positive; traversing *with* current makes it negative.

Kirchoff's Laws–Using the Approach

#### Example 8: Determine the meter

*reading* in the circuit to the right using Kirchoff's Laws. Assume the power supply is ideal with an EMF of 10 volts, and assume the resistor values are the same as their subscripts (this is essentially *Example 4*).

Step o: Remove the meters.

Step 1: Define one current for eachbranch.<br/>Step 2: Write out node equations for as<br/>many nodes as you can (see note below). Be sure to identify<br/>which node you are working with. For this problem: Node A:

*Important note*: If you had written out the node equation for the node at the bottom, you would have gotten  $i_2 + i_3 = i_0$ . This is the same equation as above. There will always be *fewer independent node equations* than actual nodes in a circuit. In this case, there were two nodes and only one independent node equation.



 $i_0 = i_2 + i_3$ 

*Additional note:* You have three branches and three unknown currents, which means you will need *three equations to solve*. You have one *node equation*, which means you will need two more equations, presumably from your *loops*. Kindly note: there are three loops in this circuit, but you can only get TWO INDEPENDENT LOOP EQUATIONS



from them. Any two of those equations will do, and any two will produce the third, which means that if you try to do this problem using nothing but loop equations, you'll end up with mush. (Try it if you don't believe me!)

*Step 3.* Identify and label the *loops* you will use. Use an arrow in each to show the direction you intend to traverse that loop.

*Note 1: If there is* a power supply in the loop, I prefer to start at the low voltage terminal and proceed through the supply. That way, the voltage change through the supply will be positive. With that in mind: Loop 1:  $\epsilon - R_1 i_0 - R_2 i_2 = 0$  Loop 2:  $R_2 i_2 - R_3 i_3 - R_4 i_3 = 0$  *Note 2: Put* resistance terms first as they'll usually be assumed known whereas currents will not be.

## Solving 3 Equations with 3 Unknowns

We have three equations and three unknowns. The ammeter is in the branch whose current is  $i_0$ . So how to solve for  $i_0$ ? There are three approaches.

Our equations:

 $\epsilon - R_1 i_0 - R_2 i_2 = 0$  (equ. A)  $R_2 i_2 - R_3 i_3 - R_4 i_3 = 0$  (equ. B)  $i_0 = i_2 + i_3$  (equ. C)

Putting in the numbers to make life easier:

 $10 - i_{o} - 2i_{2} = 0 \quad (equ. A) \qquad 2i_{2} - 3i_{3} - 4i_{3} = 0 \quad (equ. B)$  $\Rightarrow 2i_{2} - 7i_{3} = 0$  $i_{o} = i_{2} + i_{3} \quad (equ. C)$ 

Approach 1—Brute force algebra:

I'll lay this out on the next page, just to convince you it's not the way to go.

#### Líke I saíd, NASTY!

$$i_{o} = i_{2} + i_{3} \quad (equ. C)$$

$$as 2i_{2} - 7i_{3} = 0 \quad (equ. B)$$

$$\Rightarrow i_{3} = \frac{2}{7}i_{2}$$

$$i_{o} = i_{2} + i_{3}$$

$$= i_{2} + \frac{2}{7}i_{2} = \frac{9}{7}i_{2}$$

$$but 10 - i_{o} - 2i_{2} = 0 \quad (equ. A)$$

$$\Rightarrow i_{2} = \frac{10 - i_{o}}{2} = \frac{10}{2} - \frac{1}{2}i_{o}$$

$$so \quad i_{o} = \frac{9}{7}i_{2} = \frac{9}{7}\left(\frac{10}{2} - \frac{1}{2}i_{0}\right)$$

$$\Rightarrow \quad i_{o} = \frac{90}{14} - \frac{9}{14}i_{0}$$

$$\Rightarrow \quad i_{o} = \frac{90}{23}$$

$$\Rightarrow \quad i_{o} = 3.91 A$$

#### Approaches 2 and 3: Matríces:

--Begin by rewriting each equation so their  $i_0$  term is in the first column, its  $i_2$  term is in the second column, etc., and its voltage term (if there is one) is on the right side of the equal sign.

Our equations become:

$$\begin{split} \epsilon - R_1 i_0 - R_2 i_2 &= 0 \quad \text{becomes} \quad R_1 i_0 + R_2 i_2 + 0 i_3 = \epsilon \\ R_2 i_2 - R_3 i_3 - R_4 i_3 &= 0 \quad \text{becomes} \quad 0 i_0 + R_2 i_2 - (R_3 + R_4) i_3 = 0 \\ i_0 &= i_2 + i_3 \quad \text{becomes} \quad i_0 - i_2 - i_3 = 0 \\ \hline --Put \ the \\ information \quad i_0 & i_2 \\ column \ column \quad column \quad column \quad voltage \\ column \\ \hline \end{pmatrix}$$

into a matrix:

 $\begin{vmatrix} i_{0} & i_{2} & i_{3} \\ column & column & column \end{vmatrix} \quad \begin{cases} voltage \\ column \\ voltage \\ column \end{vmatrix}$   $\begin{vmatrix} R_{1} & R_{2} & 0 \\ 0 & R_{2} & -(R_{3} + R_{4}) \\ 1 & -1 & -1 \end{vmatrix} \quad \begin{vmatrix} i_{0} \\ i_{2} \\ i_{3} \end{vmatrix} = \begin{vmatrix} \varepsilon \\ 0 \\ 0 \end{vmatrix}$ 

#### --Usíng numbers:

$$\begin{vmatrix} R_1 & R_2 & 0 \\ 0 & R_2 & -(R_3 + R_4) \\ 1 & -1 & -1 \end{vmatrix} \begin{vmatrix} i_0 \\ i_2 \\ i_3 \end{vmatrix} = \begin{vmatrix} \varepsilon \\ 0 \\ 0 \end{vmatrix} \text{ becomes } \begin{vmatrix} 1 & 2 & 0 \\ 0 & 2 & -7 \\ 1 & -1 & -1 \end{vmatrix} \begin{vmatrix} i_0 \\ i_2 \\ i_3 \end{vmatrix} = \begin{vmatrix} 10 \\ 0 \\ 0 \end{vmatrix}$$

*--You have* two options at this point, depending upon your abilities with a calculator and whether there are any variables in your relationship. The first approach is a manual evaluation of the matrices and will always work.

Noting that the left-hand 3x3 matrix is called *the determinate*, solving for, say,  $i_o$ , requires the evaluation of two matrices, one divided into the other. Specifically, the *determinate* divided into the determinate with the coliumn replaced by the voltage column (the far column to the right). That is:

$$\mathbf{i}_{o} = \frac{\begin{vmatrix} 10 & 2 & 0 \\ 0 & 2 & -7 \\ 0 & -1 & -1 \end{vmatrix}}{\begin{vmatrix} 1 & 2 & 0 \\ 0 & 2 & -7 \\ 1 & -1 & -1 \end{vmatrix}}$$

--How to evaluate a matrix? Start by reproducing the first two columns at the end of the matrix.

$$\mathbf{i}_{0} = \frac{\begin{vmatrix} \varepsilon & R_{2} & 0 \\ 0 & R_{2} & -(R_{3} + R_{4}) \\ 0 & -1 & -1 \end{vmatrix}}{\begin{vmatrix} \varepsilon & R_{2} \\ 0 & R_{2} \\ 0 & -1 \end{vmatrix}} \frac{\begin{vmatrix} \varepsilon & R_{2} \\ 0 & R_{2} \\ 0 & -1 \\ \hline R_{1} & R_{2} \\ 0 & -1 \\ \hline R_{1} & R_{2} \\ 0 & R_{2} \\ 1 & -1 & -1 \end{vmatrix}}$$

--With numbers:

$$i_{0} = \frac{\begin{vmatrix} 10 & 2 & 0 & | & 10 & 2 \\ 0 & 2 & -7 & | & 0 & 2 \\ 0 & -1 & -1 & | & 0 & -1 \\ 0 & 2 & -7 & | & 0 & 2 \\ 0 & -1 & -1 & | & 0 & -1 \\ 0 & 2 & -7 & | & 0 & 2 \\ 0 & -1 & -1 & | & 1 & -1 \end{vmatrix} = \frac{(10)[(2)(-1)-(-7)(-1)] + ...}{\text{etc.}}$$
  

$$i_{0} = \frac{\begin{vmatrix} 10 & 2 & 0 & | & 12 \\ 0 & 2 & -7 & | & 0 & 2 \\ 1 & -1 & -1 & | & 1 & -1 \end{vmatrix}$$
  
--The second part:  

$$i_{0} = \frac{\begin{vmatrix} 10 & 2 & 0 & | & 12 \\ 0 & 2 & -7 & | & 0 & 2 \\ 1 & -1 & -1 & | & 1 & -1 \end{vmatrix} = \frac{(10)[(2)(-1)-(-7)(-1)] + (2)[(-7)(0)-(0)(-1)] + ...}{\text{etc.}}$$

--Once you get the hang of the pattern, you can do these in your head without writing much of anything down:

$$\mathbf{i}_{0} = \frac{\begin{vmatrix} 10 & 2 & 0 & | & 10 & 2 \\ 0 & 2 & -7 & | & 0 & 2 \\ 0 & -1 & -1 & | & 0 & -1 \\ \hline 1 & 2 & 0 & | & 1 & 2 \\ 0 & 2 & -7 & | & 0 & 2 \\ 1 & -1 & -1 & | & 1 & -1 \end{vmatrix}} = \frac{(10)[(-2)-(7)]+0+0}{1[-2-(7)]+2[(-7)-0]+0} = \frac{-90}{-23} = 3.91 \text{ A}$$

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--The other alternative has to do with matrix manipulation on a calculator. Specifically, if you multiply everything by the *inverse determinate*, you end up with a 1x3 matrix whose elements are the solution for the three unknowns.

DET
$$\begin{vmatrix} 1 \\ D \\ E \\ T \end{vmatrix} \begin{vmatrix} 1_{o} \\ i_{2} \\ i_{3} \end{vmatrix} = \begin{vmatrix} V_{o} \\ V_{2} \\ V_{3} \end{vmatrix} DET\begin{vmatrix} 1 \\ D \\ E \\ V_{3} \end{vmatrix}$$
$$= 1$$
$$= 1$$
$$\Rightarrow \begin{vmatrix} i_{o} \\ i_{2} \\ i_{3} \end{vmatrix} = \begin{vmatrix} V_{o} \\ V_{2} \\ V_{3} \end{vmatrix} DET\begin{vmatrix} -1 \\ D \\ E \\ T \end{vmatrix}$$

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*--the alternate alternate* is to have your calculator execute an rref (reduce row echalon format) operation. The following is courtesy of Mr. White.

 $i_{o} + 2i_{2} + 0i_{3} = 10 \qquad 1 \ 2 \ 0 \ 10 \\ 0i_{o} + 2i_{2} - 7i_{3} = 0 \qquad \Rightarrow \qquad 0 \ 2 \ -7 \ 0 \\ i_{o} - i_{2} - i_{3} = 0 \qquad 1 \ -1 \ -1 \ 0$ 

Using your calculator:

a. Math -> Matrix -> Edit -> A (for name of matrix) . . . note that some calculators just have a "matrix" key you can use (versus starting with "math")

b. 3 [Enter] 4 [Enter] (this gives you a 3x4 matrix)

c. Enter coefficients and values into Matrix; exit, then go back to "matrix" and:

d. In "math," use "rref" A (reduced row echelon form)

e. You'll end up with 1's and the last row will give you the current values.

$$\begin{bmatrix} 1 & 2 & 0 & 10 \\ 0 & 2 & -7 & 0 \\ 1 & -1 & -1 & 0 \end{bmatrix} \implies \begin{bmatrix} 1 & 0 & 0 & 3.91 \\ 0 & 1 & 0 & 3.04 \\ 0 & 0 & 1 & .87 \end{bmatrix} \implies \mathbf{i}_{2} = 3.04 \text{ A}$$

## Example 9: Example 8 using a

clever shortcut. Again the power supply is ideal with an EMF of 10 volts, and assume the resistor values are the same as their subscripts.

Step o: Remove the meters.

Step 1: Define one *current* for each branch. And here is the clever move.

Think about it. If current  $i_0$  comes into

node A, and current  $i_2$  goes out of node A and through  $R_2$ , how much current must go through  $R_3$ ? Must be  $i_0 - i_2$ . So why not just call it that (instead of  $i_3$ )? Doing so *eliminates one unknown*, which makes the solving a lot easier.

*Consequence*: You only need to write two loop equations (you've already used the node information in defining the currents).



#### Loop 1: $\epsilon - R_1 i_0 - R_2 i_2 = 0$ $\Rightarrow i_0 + 2i_2 = 10$ Loop 2: $R_2 i_2 - R_3 (i_0 - i_2) - R_4 (i_0 - i_2) = 0$ $\Rightarrow 2i_2 - 3(i_0 - i_2) - 4(i_0 - i_2) = 0$

 $\Rightarrow$  -7i<sub>0</sub> +9i<sub>2</sub> = 0



Solvíng:

$$\Rightarrow i_{o} = \frac{\begin{vmatrix} 10 & 2 \\ 0 & 9 \end{vmatrix}}{\begin{vmatrix} 1 & 2 \\ -7 & 9 \end{vmatrix}} = \frac{90 - 0}{9 - (-14)}$$
$$= 3.91 \text{ A}$$

## Capacitors—Charging Characteristics

**Example 10:** Consider a resistor, an uncharged capacitor, a switch and a power supply all hooked in series. Note also that when the switch is thrown, the voltage across "a" and "b" is equal to both the *battery voltage* and the *sum of voltages across the resistor and capacitor*. That is:



 $V_{o} = V_{C} + V_{R}$ 

a.) At t = 0, the switch is closed. What initially happens in the circuit?

As the cap initially has no charge on its plates, it will provide no resistance to charge flow. That means no voltage drop across the capacitor with all the voltage drop  $V_o = N_C + V_R$ happen across the resistor  $= i_o R$ ... which means:  $V_o$ 



b.) What happens as time proceeds?

As the cap begins to charge, some of the voltage drop happens across the resistor and some across the capacitor leaving us with a Kirchoff expression of:

$$V_{o} - \frac{q_{\text{plates}}}{C} - iR = 0$$
  
$$\implies \frac{dq}{dt} + \frac{q_{\text{plates}}}{RC} = \frac{V_{o}}{R}$$

$$a \begin{bmatrix} C \\ + \\ + \\ V_{c} = q(t) \\ V_{c} = i(t)R \\ V_{o} \end{bmatrix} b$$

The problem? There are two different types of q in this expression. One refers to the amount of charge on one capacitor plate. The other refers to charge flowing through the circuit (current is defined as the *time rate of charge flow*). Although this won't always be the case, in this instance the rate at which

charge accumulates on the cap plates will equal the rate at which charge passes by per unit time, and we can write:

$$i = \frac{dq}{dt} = \frac{dq_{plate}}{dt}$$

This means Kirchoff's Law can be written as:

$$\frac{dq}{dt} + \frac{q_{\text{plates}}}{RC} = \frac{V_{\text{o}}}{R}$$
$$\implies \frac{dq_{\text{plate}}}{dt} + \frac{q_{\text{plates}}}{RC} = \frac{V_{\text{o}}}{R}$$

$$a \downarrow i \rightarrow 0 \downarrow V_{o}$$

$$R$$

$$R$$

$$R$$

$$R$$

$$V_{c} = q(t) / V_{R} = i(t)R$$

$$V_{c} = i(t)R$$

Note that as time proceeds toward infinity, the charge on the capacitor plates reaches maximum, all the voltage drop happens across the capacitor, current in the circuit drops to zero and there is no voltage drop across the resistor. In that case:

$$V_{o} = V_{C} + N_{R}^{0}$$
$$= \frac{Q_{max}}{C}$$
$$\Rightarrow Q_{max} = V_{o}C$$

Solving:  

$$\frac{dq}{dt} + \left(\frac{1}{RC}\right)q = \frac{V_o}{R}$$

$$\Rightarrow \frac{dq}{dt} = \left(\frac{1}{RC}\right)(V_oC - q) = \left(\frac{1}{RC}\right)(Q_{max} - q)$$

$$\Rightarrow \frac{dq}{dt} = \left(\frac{1}{RC}\right)(V_oC - q) = \left(\frac{1}{RC}\right)(Q_{max} - q)$$

$$\Rightarrow \frac{dq}{(q - Q_{max})} = -\frac{dt}{RC}$$

$$\Rightarrow \int_0^{q(t)} \frac{dq}{(q - Q_{max})} = -\int_{t=0}^t \frac{dt}{RC} \Rightarrow \ln|q - Q_{max}||_{q=0}^{q(t)} = -\frac{t}{RC}$$

$$\Rightarrow \ln|q(t) - Q_{max}| - \ln|-Q_{max}| = -\frac{t}{RC} \Rightarrow \ln(Q_{max} - q(t)) - \ln(Q_{max}) = -\frac{t}{RC}$$

$$\Rightarrow \ln\left[\frac{(Q_{max} - q(t))}{(Q_{max})}\right] = -\frac{t}{RC} \Rightarrow e^{\ln\frac{(Q_{max} - q(t))}{(Q_{max})}} = e^{-\frac{t}{RC}}$$

$$\Rightarrow \frac{(Q_{max} - q(t))}{(Q_{max})} = e^{-\frac{t}{RC}} \Rightarrow Q_{max} - q(t) = Q_{max}e^{-\frac{t}{RC}} \Rightarrow q(t) = Q_{max}\left(1 - e^{-\frac{t}{RC}}\right)$$

32.)

Time Constant for a Capacitor

A graph of the *charging* characteristic of a charging capacitor is shown below.



*It would* be nice to get a feel for how fast a capacitor/resistor combination will charge or discharge.

charge on

capacitor

To that end, how much

charge would the cap have accumulated after a time equal to *RC*?



This time is defined as one time constant  $\tau$ . It is the amount of time it takes the capacitor to charge to 63% of its maximum. *Two time constants* will charge it to 87% of its maximum (try the calculation if you don't believe me).

c.) What is the *current* as a function of time?

$$i(t) = \frac{dq_{plate}}{dt}$$

$$= \frac{d(Q_{max} - Q_{max}e^{-t/RC})}{dt}$$

$$= -Q_{max}\left(-\frac{1}{RC}\right)e^{-t/RC}$$

$$= \left(\frac{1}{R}\right)\left(\frac{Q_{max}}{C}\right)e^{-t/RC}$$

$$= \left(\frac{V_o}{R}\right)e^{-t/RC}$$

$$= i_o e^{-t/RC}$$



A graph of the *current* characteristics for a charging capacitor/resistor combination:



After one time constant, the capacitor's current will have dropped 63% and will be at 37% of its maximum. After two time constants, it will be at 13% of its maximum.

Capacitors—Discharging Characteristics

**Example 11:** At t = 0, the switch is thrown and a charged capacitor begins to discharge.

*a.) How are* current through the circuit and charge on the capacitor plates related?

When a capacitor is discharging, the *rate of change* of charge on the plate is negative (charge is leaving) and:

$$i = \frac{dq}{dt} = -\left(\frac{dq_{\text{plate}}}{dt}\right)$$

*Using this with* Kirchoff's Law (tracking along the direction of current flow) yields:





Solving: switch closed at t = 0 $-\left(-\frac{\mathrm{d}q_{\mathrm{plate}}}{\mathrm{d}t}\right) + \left(\frac{1}{\mathrm{RC}}\right)q_{\mathrm{plates}} = 0$ i(t)  $\Rightarrow \frac{dq_{\text{plate}}}{dt} = -\left(\frac{1}{RC}\right)q_{\text{plates}}$ R  $\Rightarrow \frac{\mathrm{dq}_{\mathrm{plate}}}{\mathrm{q}_{\mathrm{plate}}} = -\left(\frac{1}{\mathrm{RC}}\right)\mathrm{dt}$  $\Rightarrow \int_{Q_{max}}^{q(t)} \left(\frac{1}{\alpha_{max}}\right) dq_{plate} = -\left(\frac{1}{RC}\right) \int_{t=0}^{t} dt$  $\ln(q)\Big|_{Q_{\text{max}}}^{q(t)} = \ln[q(t)) - \ln(Q_{\text{max}})\Big] = -\frac{t}{\mathbf{PC}}$  $\Rightarrow \ln\left(\frac{q(t)}{Q_{max}}\right) = -\frac{t}{RC} \Rightarrow e^{\ln\left(\frac{q(t)}{Q_{max}}\right)} = e^{-\frac{t}{RC}}$  $\Rightarrow \frac{q(t)}{Q} = e^{-\frac{t}{RC}} \Rightarrow q(t) = Q_{max} e^{-\frac{t}{RC}}$ 

A graph of the *charge on plate* characteristics for a discharging capacitor/resistor combination:



After one time constant, the capacitor's charge will have dropped 67% and will be at 37% of its maximum. After two time constants, it will be at 13% of its maximum.

c.) What is the *current* as a function of time?

$$i(t) = -\frac{dq(t)_{plate}}{dt}$$
$$= -\frac{d\left(Q_{max}e^{-\frac{t}{RC}}\right)}{dt}$$
$$= -\left(-\frac{1}{RC}\right)Q_{max}e^{-\frac{t}{RC}}$$
$$= \frac{1}{R}\left(\frac{Q_{max}}{C}\right)e^{-\frac{t}{RC}}$$
$$= \frac{1}{R}(V_{o})e^{-\frac{t}{RC}}$$
$$= i_{o}e^{-\frac{t}{RC}}$$



# A graph of the *current* characteristics for a discharging capacitor/resistor combination:



After one time constant, the capacitor's current will have dropped 67% and will be at 37% of its maximum. After two time constants, it will be at 13% of its maximum.

And yes, it's the same as for a charging circuit.



**Example 11:** A one microfarad cap is in series with a 10 k-ohm resistor, a battery whose voltage is 100 volts and a switch.

*a.) The capacitance value* tells you something that is always true no matter what the voltage across the cap happens to be. What does it tell you?

The capacitance tells you the amount of charge one plate can hold per volt across the plates.

b.) What is the initial current in the circuit?

*With no charge* initial on the cap (and, hence, no voltage across the cap), we can write:

c.) What is the circuit's current after a long time? It will go to zero.

d.) How much charge will the cap hold when fully charged?  $V_o = V_{cap} + V_{res}$ When fully charged, all the voltage drop will be across the cap (no current in the circuit, so no drop across the resistor), and  $V_o = V_{cap} + V_{res}$   $(100 \text{ V}) = \frac{Q_{max}}{C} + 0$   $(100 \text{ V}) = \frac{Q_{max}}{C} + 0$  $(100 \text{ V}) = \frac{Q_{max}}{C} + 0$ 



 $V_{o} = V_{cap} + V_{res}$ (100 V) = 0 + i<sub>o</sub>(10<sup>4</sup> Ω)  $\Rightarrow i_{o} = 10^{-2} A$ 

44.



*g.)* You are told the system's time constant is  $10^{-2}$  seconds. What does that tell you? It tells you how long it will take for the cap to charge up to or discharge by 63%.

*h.*) Where is the charge alluded to found?On one plate of the cap.

#### THINGS TO NOTICE ABOUT CAPACITORS

- 1.) Caps in series have common charge on their plates.
- 2.) Caps in parallel have common voltage across their plates.

3.) Close  $S_1$  with  $S_2$  open and  $C_1$  charges to Q. Keep  $S_1$  closed and close  $S_2$ .  $C_1$  is still across the battery so it keeps Q on it, but  $C_2$  is now also across the battery so more charge is drawn from the battery. Because the battery is still connected, *the total charge in the system is not fixed*.

4.) Different scenario: Close  $S_1$  with  $S_2$  open and  $C_1$  charges to Q. Open  $S_1$  and  $C_1$  is now isolated (disconnected from the battery). Close  $S_2$  so that  $C_1$  and  $C_2$  are now in parallel. Because the battery is no longer in the circuit, the charge in the system is fixed. That means the charge on  $C_1$  has to redistribute itself which it will do until the voltage across each cap is the same (and if the charge that flows from  $C_1$  to  $C_2$  is q, then the charge left on  $C_1$  is (Q - q)).

5.) Second circuit: S starts out closed so  $C_1$  charges to Q. S is opened so the caps are now in series. What changes? Because the battery voltage is now distributed between the two caps,  $C_1$ 's voltage goes down but the charge on the two caps has to be the same.





**Example 12:** You charge up two unequal capacitors that are in series. You disconnect the battery by opening  $S_1$ , then reconnect the two caps by closing  $S_2$ .

a.) What is initially common to the two caps? Caps in series have the same amount of charge on their plates.

*b.) When you* throw both switches, how are the caps related (series or parallel)?

They are now connected in two places, which is characteristic of a parallel combination.

c.) So what should happen when both switches are thrown?

*The the caps* now in parallel, their voltages have to be the same so charge will have to rearrange itself necessitating current to momentarily flow.



**Example 11:** (courtesy of Mr. White): Examine the circuit here, where  $C_1 > C_2$  and both have been charged to the same potential *V*.

*a.) What is the* potential between points *a* across *b* before the switches are closed? Q1/C1



b.) What happens to the charges after the switches are thrown?

*This is a screwball problem* in the sense that the voltages initially have the same magnitude, but their polarities are reversed. That means that when the switches are thrown, charge will flow between the caps. Specifically:

 $Q_1 > Q_2$  (as  $C_1 > C_2$ ), so excess  $Q_1$  will flow until  $V_1 = V_2$  again (but now with the right polarity)

*c.)* What is the potential across *a* and *b* a long time after the switches have been closed?

This is going to take some room:

c.) What is the potential across a and b a long time after the switches have been closed?

The total charge in the system is:  $Q_1 - Q_2 = C_1 V - C_2 V$ 

Letting the new charge on  $C_1$  be q, the new charge on  $C_2$ will be:  $(Q_1 - Q_2) - q = (C_1 V - C_2 V) - q$ 

With the new voltages the same, we can write:

 $V_{C1} = V_{C2}$   $\frac{q}{C_{1}} = \frac{(C_{1} - C_{2})V - q}{C_{2}}$   $\Rightarrow qC_{2} = C_{1}(C_{1} - C_{2})V - C_{1}q$   $\Rightarrow qC_{2} + qC_{1} = C_{1}(C_{1} - C_{2})V$   $\Rightarrow q(C_{1} + C_{2}) = C_{1}(C_{1} - C_{2})V$   $\Rightarrow q = \frac{C_{1}(C_{1} - C_{2})V}{(C_{1} + C_{2})}$ 

Knowing q, we can write:

te:  

$$V_{C1} = \frac{q}{C_1} = \frac{Q_1'(C_1 - C_2)V}{Q_1'} = \frac{(C_1 - C_2)V}{(C_1 + C_2)}$$



*d.*) *What is the* energy stored in the system before and after the closing of the switches?

$$U_{initial} = \frac{1}{2}C_1V^2 + \frac{1}{2}C_2V^2 = \frac{1}{2}(C_1 + C_2)V^2$$
$$U_{final} = \frac{1}{2}C_{equivalent}V'^2 = \frac{1}{2}(C_1 + C_2)\left(V\frac{C_1 - C_2}{C_1 + C_2}\right)^2$$
$$\left(C_1 - C_2\right)^2$$

$$U_{final} = \left(\frac{C_1 - C_2}{C_1 + C_2}\right)^2 U_{initial}$$



The Galvanometer

A galvanometer is an ammeter built to a very specific specification. All galvanometers swing maximum deflection when  $5 \times 10^{-4}$  amps flow through them. Although they are usually discussed in the chapter on magnetism (they are built using a coil in a magnetic field), it is possible to use a galvanometer to build an ammeter designed to handle larger currents, and to build voltmeters.







**Example 12: Design** a 2-amp ammeter assuming the galvanometer's resistance is 12  $\Omega$ . *For the galvanometer* to go full deflection--something you want it to do when 2 amps flow into it--you need to shunt some of the current away from the galvanometer. The current that does not flow into the galvanometer flows through the shunt resistor. In this case, that will be:

2 - .0005 = 1.9995 amps.

With the shunt resistor in parallel with the galvanometer and the voltage across each being the same, we can write:

$$i_{g}R_{g} = i_{s}R_{s}$$
  

$$\Rightarrow (5x10^{-4} A)(12 \Omega) = 1.9995(R)$$
  

$$\Rightarrow R \approx 3x10^{-3} \Omega$$



**Example 13:** Design a 12-volt voltmeter assuming the galvanometer's resistance is 12  $\Omega$ .

You want the galvanometer to go full deflection when 12 volts is across its terminals. Unfortunately, 12 volts will produce a HUGE current, so you need to cut the down current. You can do that by adding a resistor in series. As the current in a series combination is the same everywhere, we can write:



$$V = i_g R_g + i_g R$$
  

$$\Rightarrow 2 = (5x10^{-4} A)(12 \Omega) + (5x10^{-4} A)(R)$$
  

$$\Rightarrow R \approx 4x10^3 \Omega$$

#### **Observations:**

--The galvanometer-engineered ammeter consists of a 12  $\Omega$  galvanometer in parallel with (in this case) a  $3x10^{-3}\Omega$  resistor (that is, essentially a wire). As the equivalent resistance of a parallel combination is *smaller than the smallest resistor* in the combination, that means that the equivalent resistance of the ammeter is REALLY SMALL—exactly as expect.

--The galvanometer-engineered voltmeter consists of galvanometer and, in this case, an additional  $40,000 \ \Omega$ resistor in series. As the equivalent resistance of a series combination is *larger than the largest resistor* in the combination, that means that the equivalent resistance of the voltmeter is REALLY Big—again, exactly as expect.



